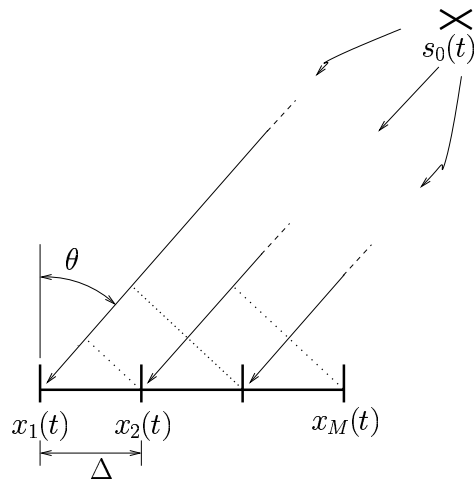


5. DIRECTION ESTIMATION USING ESPRIT

Problem

From the output of a **uniform linear antenna array**, estimate the angles of arrival



$$\mathbf{X} = \mathbf{A}\mathbf{S} = \mathbf{a}(\theta_1)\beta_1\mathbf{s}_1 + \mathbf{a}(\theta_2)\beta_2\mathbf{s}_2 + \dots$$

- General parametric procedure for DOA estimation:

$$\arg \min_{\{\theta_i\}} \|\mathbf{X} - \mathbf{A}\mathbf{S}\|_F^2, \quad \text{where } \mathbf{A} = [\mathbf{a}(\theta_1) \ \mathbf{a}(\theta_2) \ \dots]$$

- We have to know the array manifold $\mathbf{a}(\theta)$.

The ESPRIT algorithm

- For a uniform linear array, $\phi = e^{j\Delta 2\pi \sin(\theta)}$, and thus

$$\mathbf{a}(\theta) = \begin{bmatrix} 1 \\ \phi \\ \phi^2 \\ \vdots \\ \phi^{M-1} \end{bmatrix} \quad \left. \begin{array}{c} \text{ } \\ \text{ } \\ \text{ } \\ \text{ } \\ \text{ } \end{array} \right\} \begin{array}{c} \mathbf{a}_x(\theta) \\ \mathbf{a}_y(\theta) \end{array}$$

- Shift-invariance property:

$$\mathbf{a}_x(\theta) := \begin{bmatrix} 1 \\ \phi \\ \vdots \\ \phi^{M-2} \end{bmatrix}, \quad \mathbf{a}_y(\theta) := \begin{bmatrix} \phi \\ \phi^2 \\ \vdots \\ \phi^{M-1} \end{bmatrix}, \quad \text{so that} \quad \mathbf{a}_y(\theta) = \mathbf{a}_x(\theta)\phi$$

The ESPRIT algorithm

- Let us group the first and last $M - 1$ antennas

$$\mathbf{x}(t) = \begin{bmatrix} x_1(t) \\ \vdots \\ x_{M-1}(t) \end{bmatrix}, \quad \mathbf{y}(t) = \begin{bmatrix} x_2(t) \\ \vdots \\ x_M(t) \end{bmatrix}, \quad \begin{aligned} \mathbf{X} &= [\mathbf{x}(0) \quad \cdots \quad \mathbf{x}(N-1)] \\ \mathbf{Y} &= [\mathbf{y}(0) \quad \cdots \quad \mathbf{y}(N-1)] \end{aligned}$$

- From the shift-invariance property:

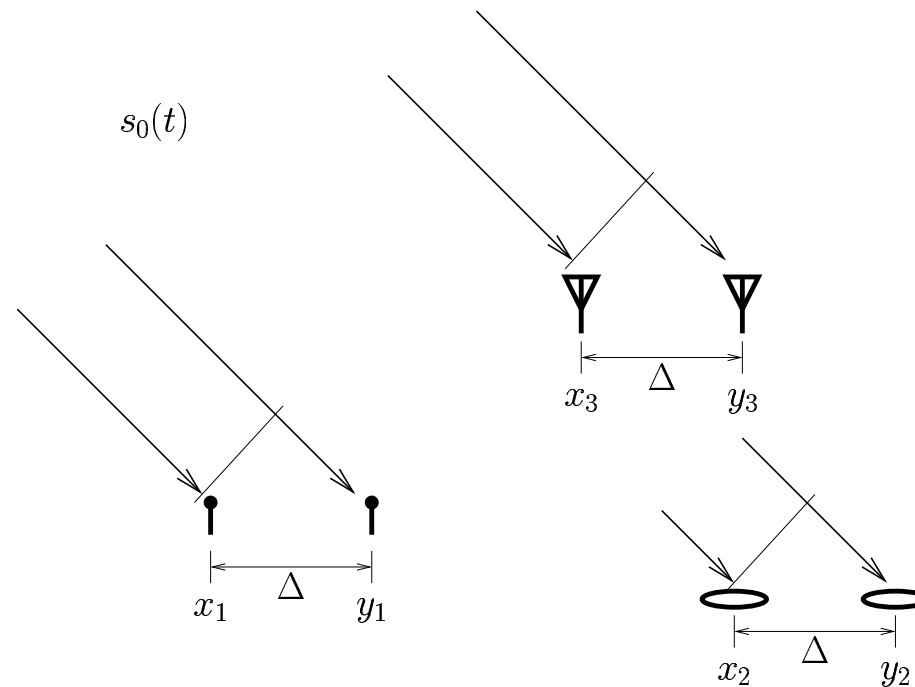
$$\begin{aligned} \mathbf{x}(t) &= \sum_{k=1}^d \mathbf{a}_x(\theta_k) \beta_k s_k(t) & \Rightarrow \quad \mathbf{X} &= \mathbf{A} \mathbf{B} \mathbf{S} \\ \mathbf{y}(t) &= \sum_{k=1}^d \mathbf{a}_y(\theta_k) \beta_k s_k(t) = \sum_{k=1}^d \mathbf{a}_x(\theta_k) \phi_k \beta_k s_k(t) & \Rightarrow \quad \mathbf{Y} &= \mathbf{A} \mathbf{\Theta} \mathbf{B} \mathbf{S} \end{aligned}$$

where

$$\mathbf{A} = [\mathbf{a}_x(\theta_1) \quad \cdots \quad \mathbf{a}_x(\theta_d)], \quad \mathbf{\Theta} = \begin{bmatrix} \phi_1 & & \\ & \ddots & \\ & & \phi_d \end{bmatrix}, \quad \phi_k = e^{j2\pi\Delta \sin(\theta_k)}$$

The ESPRIT algorithm

More general “doublet” antenna structure



$$x_i(t) = \sum_{k=1}^d a_{i,k} \beta_k s_k(t), \quad y_i(t) = \sum_{k=1}^d a_{i,k} e^{j2\pi \Delta \sin(\theta_k)} \beta_k s_k(t)$$

$$\mathbf{z}(t) = \begin{bmatrix} \mathbf{x}(t) \\ \mathbf{y}(t) \end{bmatrix} \Rightarrow \mathbf{Z} = \begin{bmatrix} \mathbf{X} \\ \mathbf{Y} \end{bmatrix} = \begin{bmatrix} \mathbf{A} \\ \mathbf{A}\Theta \end{bmatrix} \mathbf{BS}$$

The ESPRIT algorithm

Given the data matrix \mathbf{Z} from all antennas

$$\mathbf{Z} = \begin{bmatrix} \mathbf{X} \\ \mathbf{Y} \end{bmatrix} = \mathbf{A}_z \mathbf{B} \mathbf{S}, \quad \mathbf{A}_z = \begin{bmatrix} \mathbf{A} \\ \mathbf{A}\Theta \end{bmatrix}$$

- Note that \mathbf{Z} has rank d . Compute an SVD of \mathbf{Z} :

$$\mathbf{Z} = \hat{\mathbf{U}}_z \hat{\Sigma}_z \hat{\mathbf{V}}_z^H$$

$\hat{\mathbf{U}}_z : 2M \times d$ has d columns which together span the column space of \mathbf{Z}

- Note that $\hat{\mathbf{U}}_z$ spans the same space as \mathbf{A}_z , hence there exists a $d \times d$ matrix \mathbf{T} :

$$\hat{\mathbf{U}}_z = \mathbf{A}_z \mathbf{T} = \begin{bmatrix} \mathbf{A} \mathbf{T} \\ \mathbf{A} \Theta \mathbf{T} \end{bmatrix}$$

- Split \mathbf{Z} into \mathbf{X} and \mathbf{Y} , and $\hat{\mathbf{U}}_z$ accordingly into $\hat{\mathbf{U}}_x$ and $\hat{\mathbf{U}}_y$, then

$$\begin{cases} \hat{\mathbf{U}}_x &= \mathbf{A} \mathbf{T} \\ \hat{\mathbf{U}}_y &= \mathbf{A} \Theta \mathbf{T} \end{cases}$$

The ESPRIT algorithm

$$\begin{cases} \hat{\mathbf{U}}_x &= \mathbf{A}\mathbf{T} \\ \hat{\mathbf{U}}_y &= \mathbf{A}\mathbf{\Theta}\mathbf{T} \end{cases}$$

- Note that $\hat{\mathbf{U}}_x^\dagger = (\mathbf{T}^H \mathbf{A}^H \mathbf{A} \mathbf{T})^{-1} \mathbf{T}^H \mathbf{A}^H = \mathbf{T}^{-1} \mathbf{A}^\dagger$ so that

$$\hat{\mathbf{U}}_x^\dagger \hat{\mathbf{U}}_y = \mathbf{T}^{-1} \mathbf{\Theta} \mathbf{T}.$$

- Thus, \mathbf{T}^{-1} and $\mathbf{\Theta}$ are given by the eigenvectors and eigenvalues of $\hat{\mathbf{U}}_x^\dagger \hat{\mathbf{U}}_y$.
- From $\mathbf{\Theta}$ we can derive $\{\phi_k\}$ and hence $\{\theta_k\}$
- From \mathbf{T} we can derive a zero-forcing beamformer on \mathbf{Z} as

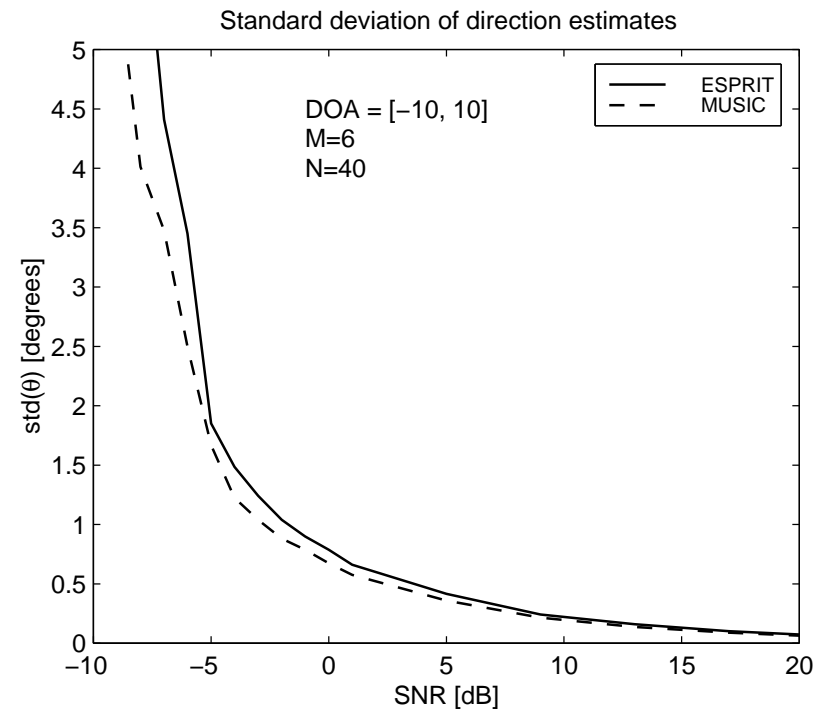
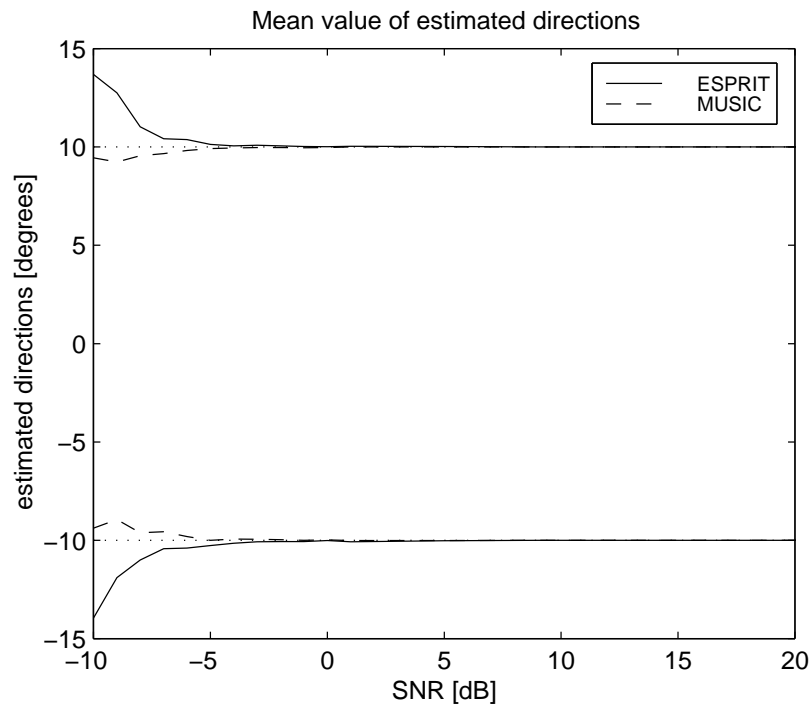
$$\mathbf{W} = \hat{\mathbf{U}}_z \mathbf{T}^H$$

Proof:

$$\begin{aligned} \mathbf{Z} &= \hat{\mathbf{U}}_z \cdot \hat{\Sigma}_z \hat{\mathbf{V}}_z^H = \mathbf{A}_z \mathbf{S} = \mathbf{A}_z \mathbf{T} \cdot \mathbf{T}^{-1} \mathbf{S} \\ \Rightarrow \quad \mathbf{T}^{-1} \mathbf{S} &= \hat{\Sigma}_z \hat{\mathbf{V}}_z^H = \hat{\mathbf{U}}_z^H \mathbf{Z} \\ \Rightarrow \quad \mathbf{S} &= \mathbf{T} \hat{\mathbf{U}}_z^H \mathbf{Z} \end{aligned}$$

The ESPRIT algorithm

Performance (varying SNR)

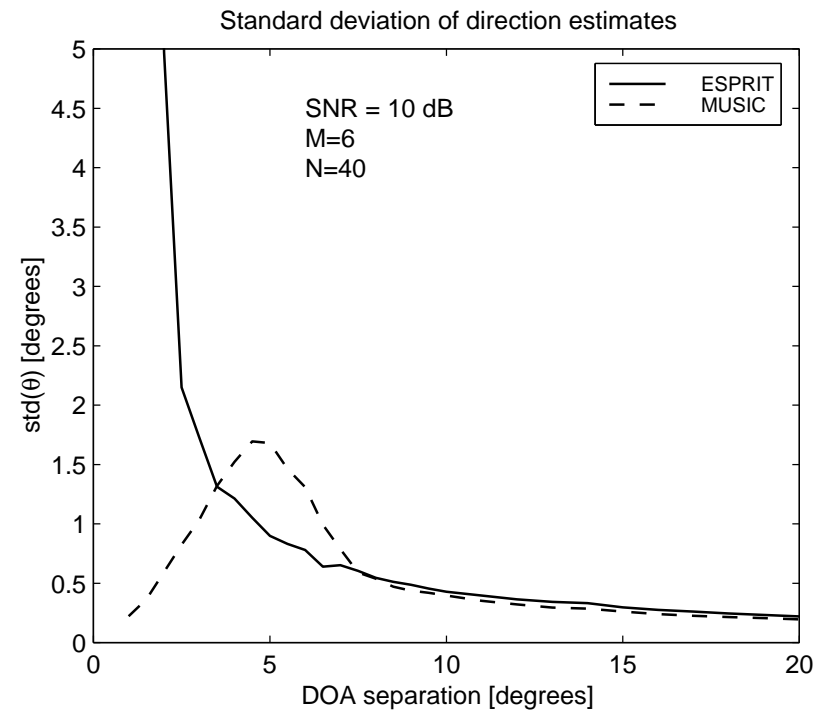
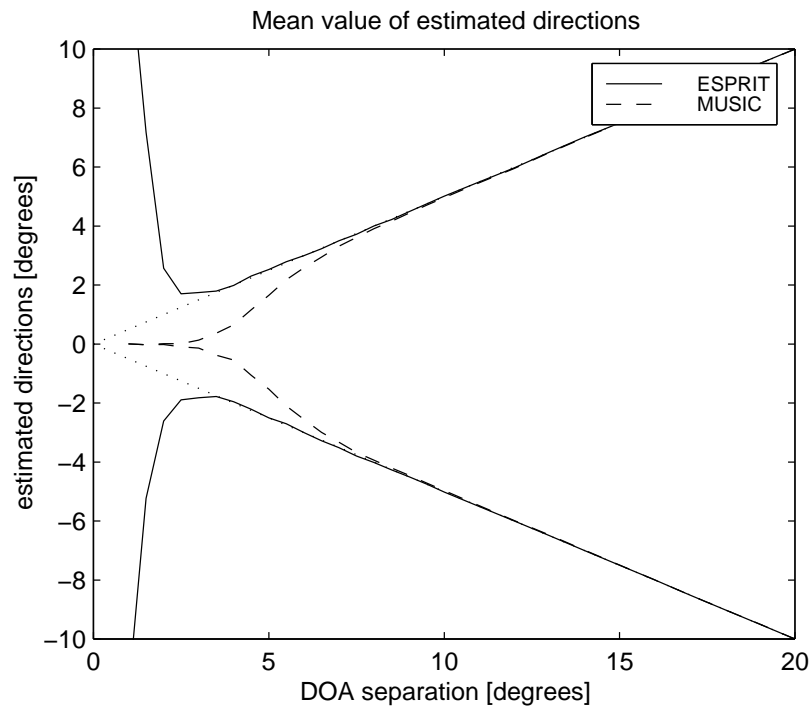


$d = 2$ sources, $M = 6$ antennas, $N = 40$ samples, 20° separation.

Mean and standard deviations of ESPRIT and MUSIC as function of SNR

The ESPRIT algorithm

Performance (varying separation)



Mean and standard deviations of ESPRIT and MUSIC as function of DOA separation.

Delay estimation using ESPRIT

Principle

- Consider an FIR pulse shape function $g(t)$ that is oversampled by a factor P :

$$g(t) \leftrightarrow \mathbf{g}(0) = \begin{bmatrix} g(0) \\ g(\frac{1}{P}) \\ \vdots \\ g(L - \frac{1}{P}) \end{bmatrix}$$

$$g(t - \tau) \leftrightarrow \mathbf{g}(\tau) = \begin{bmatrix} g(0 - \tau) \\ g(\frac{1}{P} - \tau) \\ \vdots \\ g(L - \frac{1}{P} - \tau) \end{bmatrix}$$

- Q: Given $\mathbf{g}(\tau)$ and knowing $\mathbf{g}(0)$, how do we estimate τ ?

A: By using the fact that a Fourier transform maps a delay to a phase shift.

Delay estimation using ESPRIT

- Apply the DFT to $\mathbf{g}(0)$:

$$\tilde{\mathbf{g}}(0) := \mathcal{F} \mathbf{g}(0), \quad \mathcal{F} := \begin{bmatrix} 1 & 1 & \dots & 1 \\ 1 & W & \dots & W^{LP-1} \\ \vdots & \vdots & & \vdots \\ 1 & W^{LP-1} & \dots & W^{(LP-1)^2} \end{bmatrix}, \quad W = e^{-j \frac{2\pi}{LP}}$$

- The DFT of $\mathbf{g}(\tau)$ can then be written as

$$\tilde{\mathbf{g}}(\tau) := \mathcal{F} \mathbf{g}(\tau) = \tilde{\mathbf{g}}(0) \odot \begin{bmatrix} 1 \\ W^{\tau P} \\ (W^{\tau P})^2 \\ \vdots \\ (W^{\tau P})^{LP-1} \end{bmatrix} = \text{diag}(\tilde{\mathbf{g}}(0)) \cdot \begin{bmatrix} 1 \\ W^{\tau P} \\ (W^{\tau P})^2 \\ \vdots \\ (W^{\tau P})^{LP-1} \end{bmatrix}$$

(\odot represents entrywise multiplication of the two vectors)

Delay estimation using ESPRIT

- From the DFT of $\mathbf{g}(0)$ and $\mathbf{g}(\tau)$ we can compute

$$\mathbf{z} := \{\text{diag}(\tilde{\mathbf{g}}(0))\}^{-1} \tilde{\mathbf{g}}(\tau)$$

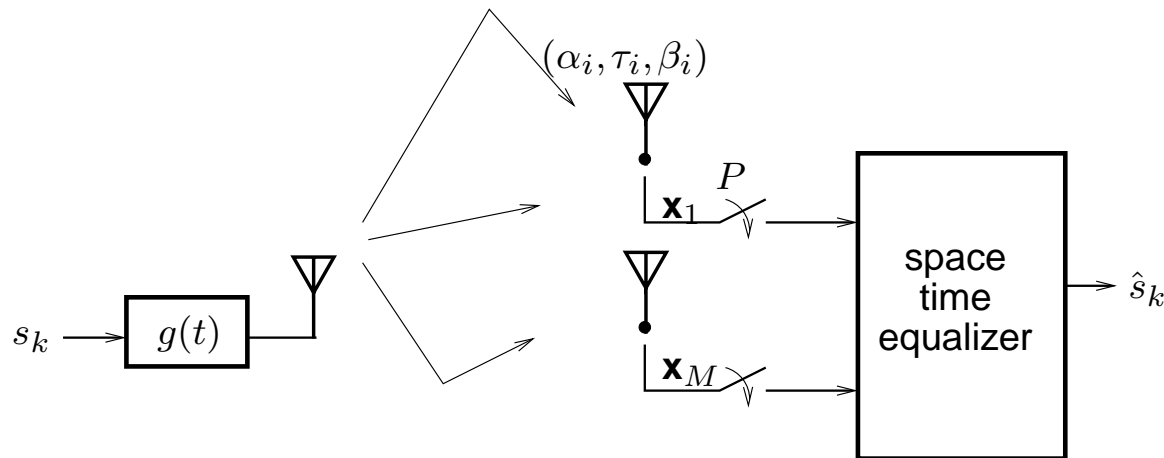
- The vector \mathbf{z} has model

$$\mathbf{z} = \mathbf{f}(\phi), \quad \mathbf{f}(\phi) := \begin{bmatrix} 1 \\ \phi \\ \phi^2 \\ \vdots \\ \phi^{LP-1} \end{bmatrix}, \quad \phi := e^{j2\pi\tau/L}$$

- Now apply ESPRIT to compute ϕ and then τ .

Delay estimation using ESPRIT

Multiple paths



- Consider a multipath channel which consists of r delayed copies of $g(t)$

$$h(t) = \sum_{i=1}^r \beta_i g(t - \tau_i) \quad \Leftrightarrow \quad \mathbf{h} = \sum_{i=1}^r \mathbf{g}(\tau_i) \beta_i = [\mathbf{g}(\tau_1), \dots, \mathbf{g}(\tau_r)] \begin{bmatrix} \beta_1 \\ \vdots \\ \beta_r \end{bmatrix} =: \mathbf{G} \boldsymbol{\beta}$$

- Assume pulse shape $g(t)$ is known and \mathbf{h} has been estimated using training.

Delay estimation using ESPRIT

- As before:

$$\mathbf{z} := \{\text{diag}(\tilde{\mathbf{g}}(0))\}^{-1} \mathcal{F} \mathbf{h}$$

- The vector \mathbf{z} has model

$$\mathbf{z} = \mathbf{F} \boldsymbol{\beta}, \quad \mathbf{F} = [\mathbf{f}(\phi_1), \dots, \mathbf{f}(\phi_r)], \quad \mathbf{f}(\phi_i) := \begin{bmatrix} 1 \\ \phi_i \\ \phi_i^2 \\ \vdots \\ \phi_i^{LP-1} \end{bmatrix}, \quad \phi_i := e^{j2\pi\tau_i/L}$$

- Consider a shift of \mathbf{z} and $\mathbf{f}(\phi_i)$ of length $LP - m + 1$:

$$\mathbf{z}^{(l)} := \begin{bmatrix} z_l \\ z_{l+1} \\ \vdots \\ z_{l+LP-m} \end{bmatrix} \quad \mathbf{f}(\phi_i)^{(l)} = \begin{bmatrix} \phi_i^l \\ \phi_i^{l+1} \\ \vdots \\ \phi_i^{l+LP-m} \end{bmatrix} = \begin{bmatrix} 1 \\ \phi_i \\ \vdots \\ \phi_i^{LP-m} \end{bmatrix} \quad \phi_i^l =: \mathbf{f}'(\phi_i) \phi_i^l$$

Delay estimation using ESPRIT

- Let us now stack the different shifts of \mathbf{z} :

$$\mathbf{Z} = [\mathbf{z}^{(0)}, \mathbf{z}^{(1)}, \dots, \mathbf{z}^{(m-1)}] = \mathbf{F}'\mathbf{B}', \quad \mathbf{F}' = [\mathbf{f}'(\phi_1), \dots, \mathbf{f}'(\phi_r)]$$

$$\mathbf{B}' = [\beta \quad \Phi\beta \quad \Phi^2\beta \quad \dots \quad \Phi^{m-1}\beta], \quad \Phi = \begin{bmatrix} \phi_1 & & \\ & \ddots & \\ & & \phi_r \end{bmatrix}$$

- Model: $\mathbf{X} = \mathbf{F}''\mathbf{B}'$, $\mathbf{Y} = \mathbf{F}''\Phi\mathbf{B}'$.

- Now we can apply ESPRIT to estimate Φ and hence all $\{\tau_i\}$.

1. Compute the SVD: $\mathbf{Z} = \hat{\mathbf{U}}_z \hat{\Sigma}_z \hat{\mathbf{V}}_z^H$

2. Split $\hat{\mathbf{U}}_z$ into $\hat{\mathbf{U}}_x$ and $\hat{\mathbf{U}}_y$ (shift over 1 row): $\hat{\mathbf{U}}_z = \begin{bmatrix} \hat{\mathbf{U}}_x \\ *** \end{bmatrix} = \begin{bmatrix} *** \\ \hat{\mathbf{U}}_y \end{bmatrix}$

Model: $\hat{\mathbf{U}}_x = \mathbf{F}''\mathbf{T}$, $\hat{\mathbf{U}}_y = \mathbf{F}''\Phi\mathbf{T}$ for some $r \times r$ matrix \mathbf{T} .

3. Compute the eigenvalue decomposition: $\hat{\mathbf{U}}_x^\dagger \hat{\mathbf{U}}_y = \mathbf{T}^{-1}\Phi\mathbf{T}$.

Delay estimation using ESPRIT

Multiple paths, multiple antennas

- The channel vector $\mathbf{h}(t)$ with M entries can then be written as

$$\mathbf{h}(t) = \sum_{i=1}^r \mathbf{a}(\theta_i) \beta_i g(t - \tau_i)$$

- Oversampling at rate P we obtain

$$\mathbf{h} = \begin{bmatrix} \mathbf{h}(0) \\ \mathbf{h}(\frac{1}{P}) \\ \vdots \\ \mathbf{h}(L - \frac{1}{P}) \end{bmatrix} = \sum_{i=1}^r [\mathbf{g}(\tau_i) \otimes \mathbf{a}(\theta_i)] \beta_i =: [\mathbf{G} \circ \mathbf{A}] \beta$$

where

Kronecker product: $\mathbf{a} \otimes \mathbf{b} := \begin{bmatrix} a_1 \mathbf{b} \\ \vdots \\ a_N \mathbf{b} \end{bmatrix}$

Khatri-Rao product: $\mathbf{A} \circ \mathbf{B} := [\mathbf{a}_1 \otimes \mathbf{b}_1 \quad \cdots \quad \mathbf{a}_r \otimes \mathbf{b}_r]$

Delay estimation using ESPRIT

- Preprocess using DFT as before and construct \mathbf{Z} from block- M shifts:

$$\mathbf{Z} = [\mathbf{z}^{(0)}, \mathbf{z}^{(1)}, \dots, \mathbf{z}^{(m-1)}] \quad M(LP - m + 1) \times m$$

- \mathbf{Z} has a model

$$\mathbf{Z} = [\mathbf{F}' \circ \mathbf{A}] \mathbf{B}', \quad \mathbf{F}' = [\mathbf{f}'(\phi_1), \dots, \mathbf{f}'(\phi_r)], \quad \mathbf{f}'(\phi_i) = \begin{bmatrix} 1 \\ \phi_i \\ \vdots \\ \phi_i^{LP-m} \end{bmatrix}$$

- Note

$$\mathbf{f}'(\phi_i) \otimes \mathbf{a}(\theta_i) = \begin{bmatrix} \mathbf{a}(\theta_i) \\ \phi_i \mathbf{a}(\theta_i) \\ \vdots \\ \phi_i^{LP-m} \mathbf{a}(\theta_i) \end{bmatrix}$$

- We can now apply the ESPRIT algorithm to \mathbf{Z} , where \mathbf{X} and \mathbf{Y} are submatrices of \mathbf{Z} with last/first M rows omitted.

Frequency estimation

Use of ESPRIT to estimate frequencies

- Given that signal $x(t)$ is the sum of d harmonic components, $x(t) = \sum_{i=1}^d \beta_i e^{j\omega_i t}$, estimate ω_i and β_i .
- Collect N samples in a data matrix \mathbf{Z} with m rows:

$$\mathbf{Z} = \begin{bmatrix} x(0) & x(1) & \cdots & x(N-m-1) \\ x(1) & x(2) & \cdots & x(N-m) \\ \ddots & \ddots & & \ddots \\ x(m-1) & x(m) & \cdots & x(N-1) \end{bmatrix}$$

- \mathbf{Z} has model

$$\mathbf{Z} = \mathbf{AS} := \begin{bmatrix} 1 & \cdots & 1 \\ \phi_1 & \cdots & \phi_d \\ \phi_1^2 & \cdots & \phi_d^2 \\ \vdots & & \vdots \end{bmatrix} \begin{bmatrix} \beta_1 & \beta_1 \phi_1 & \beta_1 \phi_1^2 & \cdots \\ \vdots & \vdots & & \\ \beta_d & \beta_d \phi_d & \beta_d \phi_d^2 & \cdots \end{bmatrix}, \quad \phi_i = e^{j\omega_i}$$