#### EE2S31 Signal Processing – Stochastic Processes

Lecture 6: Filtering stochastic processes - Suppl. 1, 2

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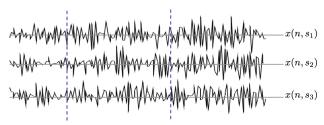
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25 May 2022



#### Ch.13.9 Ergodicity

#### Estimating expected value: ensemble average



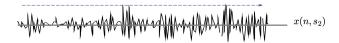
How can we estimate E[X(n)] if we don't know the PDF?

■ Ensemble average:  $\hat{\mu}(n) = \frac{1}{I} \sum_{i=1}^{I} x(n, s_i)$ .

We will need many independent observations!

■ If WSS process: E[X(n)] is the same for all n. Can we use that?

If the process is **ergodic**, we can also average over time using a single realization (in this case  $x(n, s_2)$ ):



$$\hat{\mu} = \frac{1}{N} \sum_{n=1}^{N} x(n, s_i)$$

**Definition:** for an *ergodic process*, the time average  $\bar{X}$  and the ensemble average E[X] are the same.

#### **Definition:**

For a stationary random process X(t), define the time averages of a sample function x(t) as

$$\bar{X}(T) = \frac{1}{2T} \int_{-T}^{T} x(t) dt$$

$$\overline{X^2}(T) = \frac{1}{2T} \int_{-T}^{T} x^2(t) dt$$

These can be measured from a single available observation.

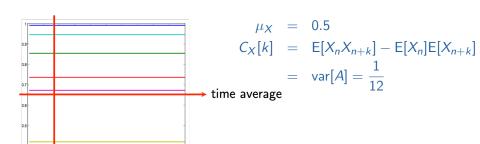
By definition, for an ergodic process

$$\lim_{T\to\infty} \bar{X}(T) = \mu_X$$

■ WSS is not sufficient! The autocovariance  $C_X(\tau)$  must go to zero quickly enough, so time samples are sufficiently independent.

#### Not all WSS processes are ergodic!

Process:  $X_n = A$ , with random amplitude A, uniform in [0,1].



ensemble average

**Theorem 13.13** Let X(t) be stationary, with expected value  $\mu_X$  and autocovariance  $C_X(\tau)$ .

If  $\int_{-\infty}^{\infty} |C_X(\tau)| d\tau < \infty$ , then the sequence  $\bar{X}(T), \bar{X}(2T), \cdots$  is an unbiased, consistent sequence of estimators of  $\mu_X$ .

It suffices that  $C_X(0) < \infty$  (finite variance) and  $C_X(\tau) = 0$  for  $\tau > \tau_0$ .

#### **Proof**

Unbiased:

$$E[\bar{X}(T)] = \frac{1}{2T} E\left[\int_{-T}^{T} X(t) dt\right] = \frac{1}{2T} \int_{-T}^{T} E[X(t)] dt$$
$$= \frac{1}{2T} \int_{-T}^{T} \mu_X dt = \mu_X$$

#### **Proof (continued)**

Consistent: sufficient to show that  $\lim_{T\to\infty} \text{var}[\bar{X}(T)] = 0$ :

$$\operatorname{var}[\bar{X}(T)] = \operatorname{E}\left[\left(\frac{1}{2T} \int_{-T}^{T} (X(t) - \mu_{X}) dt\right)^{2}\right]$$

$$= \frac{1}{(2T)^{2}} \operatorname{E}\left[\left(\int_{-T}^{T} (X(t) - \mu_{X}) dt\right) \left(\int_{-T}^{T} (X(t' - \mu_{X}) dt')\right)\right]$$

$$= \frac{1}{(2T)^{2}} \int_{-T}^{T} \int_{-T}^{T} \operatorname{E}[(X(t) - \mu_{X})(X(t') - \mu_{X})] dt' dt$$

$$= \frac{1}{(2T)^{2}} \int_{-T}^{T} \underbrace{\int_{-T}^{T} C_{X}(t - t') dt'}_{\text{bounded by some } K} dt$$

**Proof (continued)** 

Note that

$$\int_{-T}^{T} C_X(t-t') \mathrm{d}t' \leq \int_{-\infty}^{\infty} |C_X(\tau)| \mathrm{d}\tau < \infty$$

so that there is a constant K such that

$$\operatorname{var}[\bar{X}(T)] \leq \frac{1}{(2T)^2} \int_{-T}^{T} K dt = \frac{K}{2T}$$

Thus  $\lim_{T\to\infty} \operatorname{var}[\bar{X}(T)] \leq \lim_{T\to\infty} \frac{K}{2T} = 0.$ 

# Similar for the Autocorrelation Function (1)

Ensemble average: 
$$\hat{R}_X[k] = \frac{1}{I} \sum_{i=1}^I x(n,s_i)x(n+k,s_i)$$

$$x(n,s_1)$$

$$x(n,s_2)$$

$$x(n,s_3)$$

Because the process is WSS, the value of n is not important.

# Similar for the Autocorrelation Function (2)

Using time averages, the autocorrelation function can be estimated from a single observation as

$$\bar{R}_X[k] = \frac{1}{N} \sum_{n=1}^{N} x(n, s_i) x(n + k, s_i)$$



### Similar for the Autocorrelation Function (3)

The basic estimator form for time averages

$$\bar{R}_X[k] = \frac{1}{N} \sum_{n=1}^{N} x(n, s_i) x(n + k, s_i)$$

uses 2N-1 data samples to estimate N lags of  $R_K[k]$ .

■ Example for k = 0, 1, 2 and N = 3:

$$R_X[0] = \frac{1}{3} \{x(1)^2 + x(2)^2 + x(3)^2 \}$$

$$R_X[1] = \frac{1}{3} \{x(1)x(2) + x(2)x(3) + x(3)x(4) \}$$

$$R_X[2] = \frac{1}{3} \{x(1)x(3) + x(2)x(4) + x(3)x(5) \}$$

Also set  $R_X[-1] = R_X[1]$ ,  $R_X[-2] = R_X[2]$ .

## Similar for the Autocorrelation Function (4)

■ Modified estimator (using N samples to estimate N correlation lags):

$$\hat{R}_X[k] = \frac{1}{N} \sum_{n=1}^{N-k} x(n, s_i) x(n+k, s_i)$$

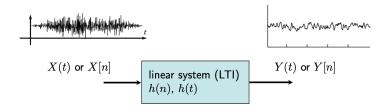
$$R_X[0] = \frac{1}{3} \{ x(1)^2 + x(2)^2 + x(3)^2 \}$$

$$R_X[1] = \frac{1}{3} \{ x(1)x(2) + x(2)x(3) \}$$

$$R_X[2] = \frac{1}{3} \{ x(1)x(3) \}$$

- This estimator is biased:  $E[\hat{R}_X[k]] = \frac{N-k}{N} R_X[k]$
- Unbiased version:  $\tilde{R}_X[k] = \frac{1}{N-k} \sum_{i=1}^{N-k} x(n, s_i) x(n+k, s_i)$

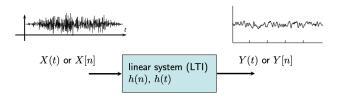
#### Suppl. 1, 2: Linear filtering of stochastic processes



Signals are often represented as sample functions of WSS processes:

- Use PDF/PMF to describe the amplitude characteristics
- Use autocorrelation to describe the time/spatial varying nature of the signals.

#### Linear filtering stochastic processes



If the input is a sample function x(t) of a random process X(t) we get

$$y(t) = \int_{-\infty}^{\infty} h(u)x(t-u)du = h(t) * x(t)$$

and therefore we write

$$Y(t) = \int_{-\infty}^{\infty} h(u)X(t-u)du = h(t) * X(t)$$

#### Expected value of the output

In general:

$$E[Y(t)] = E\left[\int_{-\infty}^{\infty} h(u)X(t-u)du\right] = \int_{-\infty}^{\infty} h(u)E[X(t-u)]du$$
$$= h(t) * E[X(t)]$$

If X(t) is WSS, then  $E[X(t)] = \mu_X$  is constant:

$$\mathsf{E}[Y(t)] = \mu_X \int_{-\infty}^{\infty} h(u) \mathrm{d}u$$

## Crosscorrelation (WSS input)

Next, we look at the autocorrelation of Y(t), and crosscorrelation of X(t) with Y(t).

It is convenient to first compute the crosscorrelation:

$$R_{XY}(\tau) = E[X(t)Y(t+\tau)]$$

$$= E\left[X(t)\int_{-\infty}^{\infty} h(v)X(t+\tau-v) dv\right]$$

$$= \int_{-\infty}^{\infty} h(v)E[X(t)X(t+\tau-v)] dv$$

$$= \int_{-\infty}^{\infty} h(v)R_X(\tau-v) dv = h(\tau)*R_X(\tau)$$

#### Autocorrelation (WSS input)

$$R_{XY}(\tau) = h(\tau) * R_X(\tau)$$

The autocorrelation of the output is then

$$R_{Y}(\tau) = \mathbb{E}[Y(t)Y(t+\tau)]$$

$$= \mathbb{E}\left[\int_{-\infty}^{\infty} h(u)X(t-u) du \int_{-\infty}^{\infty} h(v)X(t+\tau-v) dv\right]$$

$$= \int_{-\infty}^{\infty} h(u) \int_{-\infty}^{\infty} h(v) \mathbb{E}[X(t-u)X(t+\tau-v)] dvdu$$

$$= \int_{-\infty}^{\infty} h(u) \int_{-\infty}^{\infty} h(v)R_{X}(\tau-v+u) dvdu$$

$$= \int_{-\infty}^{\infty} h(u)R_{XY}(\tau+u) du = h(-\tau) * R_{XY}(\tau)$$

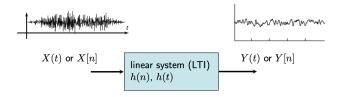
$$= h(-\tau) * h(\tau) * R_{X}(\tau)$$

#### Autocorrelation (WSS input)

Hence, if X(t) is WSS, then Y(t) is also WSS: E[Y(t)] is independent of time, and  $R_Y(t,\tau)$  only depends on the shift  $\tau$ .

Since also  $R_{XY}(t,\tau)$  only depends on  $\tau$ , we conclude that X(t) and Y(t) are jointly WSS.

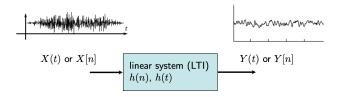
#### Output distribution



What can we say about the PDF (or PMF) of the output?

- In general this is difficult!
- Exception: a Gaussian stochastic process.

#### Output distribution

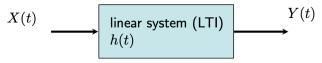


If the input X(t) is a stationary Gaussian stochastic process, and the filter is LTI with impulse response h(t),

then the output is also stationary Gaussian, with expected value and autocorrelation as specified before.

"Handwaving proof": Remember that a linear transformation of jointly Gaussian RVs gives jointly Gaussian RVs.

#### Summarizing

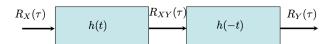


WSS input gives WSS output

Statistical descriptions of X(t): Statistical descriptions of Y(t):

- mean  $\mu_X$
- Autocorrelation  $R_X(\tau)$ .

- mean  $\mu_Y = \mu_X \int_t h(t) dt$
- $R_Y(\tau) = h(-\tau) * h(\tau) * R_X(\tau)$ .



WSS Gaussian input gives WSS Gaussian output

Let X(t) be WSS with  $\mathsf{E}[X(t)] = 10$ . Apply a linear filter with impulse response

$$h(t) = egin{cases} \mathrm{e}^{t/0.2} & 0 \leq t \leq 0.1 \ \mathrm{sec.} \ 0 & \mathrm{otherwise} \end{cases}$$

Determine E[Y(t)]

Let X(t) be WSS with  $\mathsf{E}[X(t)] = 10$ . Apply a linear filter with impulse response

$$h(t) = egin{cases} e^{t/0.2} & 0 \leq t \leq 0.1 \text{ sec.} \\ 0 & \text{otherwise} \end{cases}$$

Determine E[Y(t)]

$$\mathsf{E}[Y(t)] = \mathsf{E}[X(t)] \int_{-\infty}^{\infty} h(t) \mathrm{d}t = 10 \int_{0}^{0.1} e^{t/0.2} \mathrm{d}t = 2(e^{0.5} - 1)$$

$$W(t) \longrightarrow h(t) = \begin{cases} \frac{1}{T} & 0 \le t \le T \\ 0 & \text{otherwise} \end{cases}$$

$$Y(t)$$

$$R_W(\tau) = \eta_0 \delta(\tau)$$

Given h(t) and the white Gaussian noise process W(t) with  $R_W(\tau) = \eta_0 \, \delta(\tau)$ .

#### Find

- E[Y(t)]
- Crosscorrelation  $R_{WY}(\tau)$
- Autocorrelation  $R_Y(\tau)$

("White" means zero mean, iid)

■ E[W(t)] = 0 (white Gaussian noise process). So

$$\mathsf{E}[Y(t)] = \mathsf{E}[W(t)] \int_0^T \frac{1}{T} \mathsf{d}t = 0.$$

**Crosscorrelation of input** W(t) with output Y(t):

$$R_{WY}(\tau) = \int_{-\infty}^{\infty} h(u)R_W(\tau - u) du = \frac{\eta_0}{T} \int_0^T \delta(\tau - u) du$$
$$= \begin{cases} \frac{\eta_0}{T} & 0 \le \tau \le T \\ 0 & \text{otherwise.} \end{cases}$$

$$R_Y(\tau) = \int_{-\infty}^{\infty} h(v) R_{WY}(\tau + v) dv = \int_0^T \frac{1}{T} R_{WY}(\tau + v) dv$$

First write  $R_{WY}(\tau + v)$  as function of v:

$$R_{WY}(\tau + v) = \begin{cases} \frac{\eta_0}{T} & 0 \le \tau + v \le T \\ 0 & \text{otherwise} \end{cases} = \begin{cases} \frac{\eta_0}{T} & -\tau \le v \le T - \tau \\ 0 & \text{otherwise}. \end{cases}$$

Integration boundaries now depend on  $\tau$ . Hence, we get two cases:

$$0 \le \tau \le T \quad : \quad R_Y(\tau) = \frac{1}{T} \int_0^{T-\tau} \frac{\eta_0}{T} dv = \frac{\eta_0(T-\tau)}{T^2}$$
$$-T \le \tau \le 0 \quad : \quad R_Y(\tau) = \frac{1}{T} \int_{-\tau}^{T} \frac{\eta_0}{T} dv = \frac{\eta_0(T+\tau)}{T^2}$$

• Altogether: 
$$R_Y(\tau) = \begin{cases} \frac{\eta_0(T - |\tau|)}{T^2} & |\tau| \leq T, \\ 0 & \text{otherwise.} \end{cases}$$

$$R_X(\tau) = 4 + 3\delta(\tau)$$

$$X(t)$$

$$h(t) = \begin{cases} 3e^{-t}, & t \ge 0 \\ 0, & t < 0. \end{cases}$$

$$Y(t)$$

$$R_Y(\tau) = h(\tau) * h(-\tau) * R_X(\tau)$$
  
=  $g(\tau) * R_X(\tau)$ 

$$g(\tau) = h(\tau) * h(-\tau) = \int_{-\infty}^{\infty} 3e^{-t}u(t) 3e^{-t+\tau}u(-\tau + t) dt$$
$$= \begin{cases} 9e^{\tau} \int_{\tau}^{\infty} e^{-2t} dt = \frac{9}{2}e^{-\tau} & \text{if } \tau \ge 0\\ 9e^{\tau} \int_{0}^{\infty} e^{-2t} dt = \frac{9}{2}e^{\tau} & \text{if } \tau < 0 \end{cases}$$

$$R_X(\tau) = 4 + 3\delta(\tau)$$

$$X(t)$$

$$h(t) = \begin{cases} 3e^{-t}, & t \ge 0 \\ 0, & t < 0. \end{cases}$$

$$Y(t)$$

$$R_{Y}(\tau) = g(\tau) * R_{X}(\tau) = \left(\frac{9}{2}e^{-\tau}u(\tau) + \frac{9}{2}e^{\tau}u(-\tau)\right) * (4 + 3\delta(\tau))$$

$$= \int_{-\infty}^{+\infty} \frac{9}{2} \left(e^{-t}u(t) + e^{t}u(-t)\right) (4 + 3\delta(\tau - t))dt$$

$$= \frac{36}{2} \int_{0}^{\infty} e^{-t}dt + \frac{36}{2} \int_{-\infty}^{0} e^{t}dt + \frac{27}{2}e^{-\tau}u(\tau) + \frac{27}{2}e^{\tau}u(-\tau)$$

$$= 36 + \frac{27}{2}e^{-|\tau|}$$

### Sampling and filtering of random processes

Let X(t) be a continuous WSS process with  $E[X(t)] = \mu_X$  and  $R_X(\tau)$ .

**Sample with period**  $T_s$ :  $X_n = X(nT_s)$ . Then

$$X_n$$
 is also WSS with  $E[X_n] = \mu_X$  and  $R_X[k] = R_X(kT_s)$ , because

$$E[X_n] = E[X(nT_s)] = \mu_X$$

$$R_X[k] = E[X_nX_{n+k}] = E[X(nT_s)X([n+k]T_s)] = R_X(kT_s).$$

#### Filtering of discrete-time random sequences:

$$Y_n = h_n * X_n = \sum_j h_j X_{n-j}$$

- $\blacksquare \ \mathsf{E}[Y_n] = \mathsf{E}[X_n] \sum_j h_j$
- $R_{XY}[k] = E[X_n Y_{n+k}] = \sum_j h_j R_X[k-j] = h_k * R_X[k]$
- $R_Y[k] = E[Y_n Y_{n+k}] = \sum_i h_i \sum_i h_j R_X[k+i-j] = h_{-k} * R_{XY}[k]$

Let  $Y_n$  be a sampled version of stochastic process Y(t). Y(t) has autocorrelation function

$$R_Y( au) = egin{cases} 10^{-9} (10^{-3} - | au|) & | au| \leq 10^{-3}, \ 0 & ext{otherwise}. \end{cases}$$

What is the autocorrelation function of the sampled process  $Y_n$  if  $F_s = 10^4$  samples/sec?

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What is the autocorrelation function of the sampled process  $Y_n$  if  $F_s = 10^4$  samples/sec?

$$\begin{split} R_Y[k] &= R_Y\left(k\frac{1}{F_s}\right) = \begin{cases} 10^{-9}(10^{-3} - |k\frac{1}{F_s}|) & |k\frac{1}{F_s}| \le 10^{-3} \\ 0 & \text{otherwise} \end{cases} \\ &= \begin{cases} 10^{-9}(10^{-3} - |k \cdot 10^{-4}|) & |k \cdot 10^{-4}| \le 10^{-3}, \\ 0 & \text{otherwise}. \end{cases} \\ &= \begin{cases} 10^{-6}(1 - |0.1k|) & |k| \le 10, \\ 0 & \text{otherwise}. \end{cases} \end{split}$$

## Problem 2.7 (modified notation)

Consider  $X_n = aX_{n-1} + V_n$ , where  $V_n$ : iid,  $E[V_n] = 0$ ,  $R_V[k] = \sigma^2 \delta[k]$ .

Find  $R_X[k]$ .

#### Problem 2.7 (modified notation)

Consider  $X_n = aX_{n-1} + V_n$ , where  $V_n$ : iid,  $E[V_n] = 0$ ,  $R_V[k] = \sigma^2 \delta[k]$ .

Find  $R_X[k]$ .

$$R_{VX}[k] = E[V_{n-k}X_n] = E[V_{n-k}(aX_{n-1} + V_n)]$$

$$= aR_{VX}[k - 1] + \sigma^2 \delta[k]$$

$$\Rightarrow R_{VX}[k] = \begin{cases} \sigma^2 a^k & k \ge 0, \\ 0 & k < 0. \end{cases}$$

$$R_{XV}[k] = R_{VX}[-k]$$

$$R_X[k] = E[X_{n-k}X_n] = E[X_{n-k}(aX_{n-1} + V_n)]$$
  
=  $aR_X[k-1] + R_{XV}[k]$ 

#### Problem 2.7 (cont'd)

We saw until now:

$$R_{x}[k] = aR_{x}[k-1] + R_{xv}[k], \qquad R_{xv}[k] = \begin{cases} \sigma^{2}a^{-k} & k \leq 0, \\ 0 & k > 0. \end{cases}$$

$$k > 0$$
:  $R_X[k] = aR_X[k-1] = \dots = a^k R_X[0]$   
 $k = 0$ :  $R_X[0] = aR_X[-1] + \sigma^2 = aR_X[1] + \sigma^2 = a^2 R_X[0] + \sigma^2$   
 $R_X[0] = \frac{\sigma^2}{1 - a^2} = : \sigma_X^2$ 

It follows, for 
$$k \ge 0$$
:  $R_X[k] = a^k \sigma_X^2$   
Also, for  $k < 0$ ,  $R_X[k] = R_X[-k] = a^{-k} \sigma_X^2$   $\Rightarrow R_X[k] = a^{|k|} \frac{\sigma^2}{1 - a^2}$ 

#### To do for this lecture:

Make some selected exercises of the Supplement:

1.1, 1.3, 2.1, 2.3, 2.5, 2.7

(Unfortunately, the supplement has far fewer exercises)

Next lecture, we'll do Supplement Sections 5 and 6.