# **EE2S11: Signals and Systems**

17. Exercises for part 2 (first part)

- 1. Sampling
- 2. *z*-transform, inverse *z*-transform
- 3. DTFT (missing...)

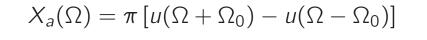
### Sampling

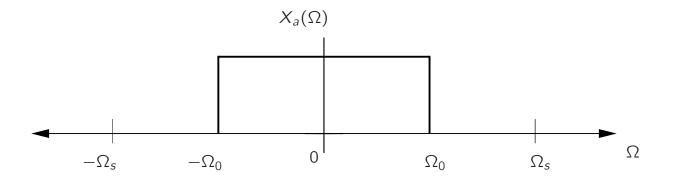
Given  $x_a(t) = \frac{\sin(\Omega_0 t)}{t}$ .

- a) What is  $X_a(\Omega)$ ? Make a drawing of  $|X_a(\Omega)|$ .
- b) Is  $x_a(t)$  band limited? (What is the maximal frequency in the signal?)
- c) At which frequency should I at least sample to avoid aliasing (i.e., the Nyquist condition)?
- d) We sample at this frequency, resulting in *x*[*n*]. Determine the corresponding spectrum *X*(ω) and make a drawing.

#### Answer

a)



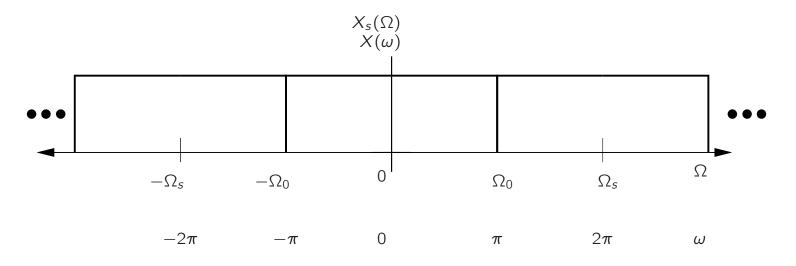


- b) Band limited: the maximal frequency is  $\Omega_0$ .
- c) Twice the highest frequency:  $\Omega_s = 2\Omega_0$ . The sample frequency is

$$F_s = \frac{\Omega_s}{2\pi} = \frac{\Omega_0}{\pi}$$

d) The spectrum of 
$$x_s(t) = \sum_n x[n]\delta(t - nT_s)$$
 is  $X_s(\Omega) = \frac{1}{T_s} \sum_k X_a(\Omega - k\Omega_s)$ .

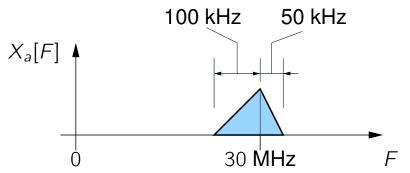
This corresponds to the spectrum  $X(\omega)$  of x[n], with  $\omega = 2\pi\Omega/\Omega_s$ , so that  $\Omega_s \to 2\pi$ .



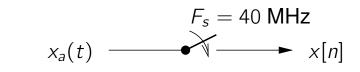
The spectrum is flat, this matches with  $x[n] = A\delta[n]$ , for a certain amplitude *A*. Check:  $x[n] = x_a(nT_s)$  with  $T_s = \frac{1}{F_s} = \frac{2\pi}{\Omega_s} = \frac{\pi}{\Omega_0}$ , so that  $x[n] = \frac{\sin(\Omega_0 \cdot n\frac{\pi}{\Omega_0})}{n\frac{\pi}{\Omega_0}} = \Omega_0 \frac{\sin(n\pi)}{n\pi} = \Omega_0 \delta[n]$ 

### Sampling

Given an analog real-valued signal  $x_a(t)$  with frequency spectrum:



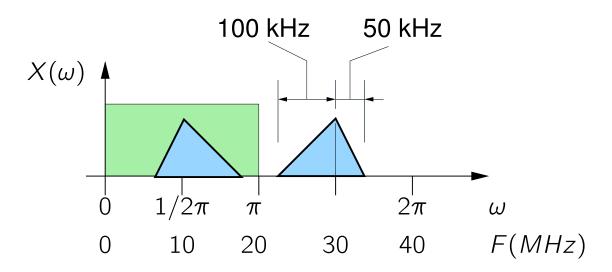
This signal is sampled with a sample frequency of 40 MHz:



- a Is the Nyquist criterion satisfied?
- b Make a drawing of the frequency spectrum of the digital signal x[n]. Also indicate which frequencies (in Hz) play a role.

#### Answer

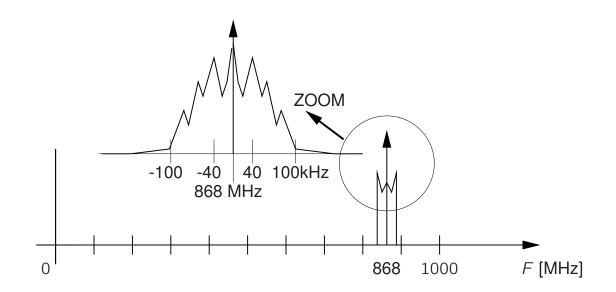
- a No: the highest frequency in the signal is 30.05 MHz, the sample frequency should be twice as large (60.1 MHz), which is not the case here.
- b Due to the sampling, part of the spectrum at 30 MHz is repeated with period 40 MHz and is seen at 70, 110 MHz etc and also at −10, −50, ··· MHz. The part of the spectrum at −30 MHz (not shown in the graph but the signal was assumed real-valued) is repeated and is seen at 10, 50, ··· MHz.



### Sampling

An RFID tag (e.g., a smart card) modulates the transmitted carrier signal of a reader device. The transmitted carrier is at 868 MHz, the modulation of the tag is a square wave with a frequency around 40 kHz.

Below, the spectrum of the received signal  $x_a(t)$  is shown (the transmitted carrier signal is dominantly present):



Make a drawing of the spectrum of x[n] if we sample at 250 MHz (accurately indicate the frequencies; also mark the fundamental interval).

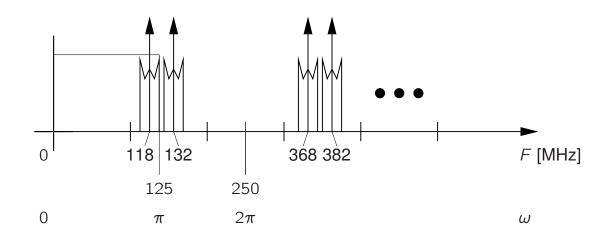
#### Answer

The spectrum becomes periodic with period 250 MHz. Due to aliasing, the frequencies around 868 MHz are shifted with multiples of 250 MHz, i.e.,

$$868 + k \cdot 250 = \begin{cases} 618 \text{ MHz}, & k = -1 \\ 368 & k = -2 \\ 118 & k = -3 \end{cases}$$

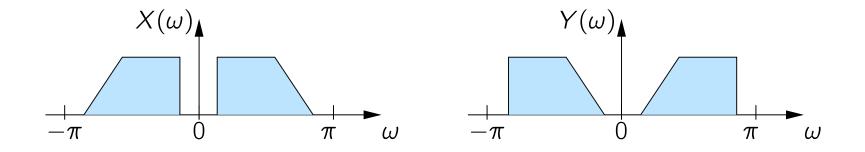
Etc. The spectrum is symmetric, we also have a part at -868 MHz which is shifted with multiples of 250 MHz, i.e.,

$$-868 + k \cdot 250 = \begin{cases} 132 \text{ MHz}, & k = 4\\ 382 & k = 5\\ 632 & k = 6 \end{cases}$$



### Spectrum of discrete-time signals

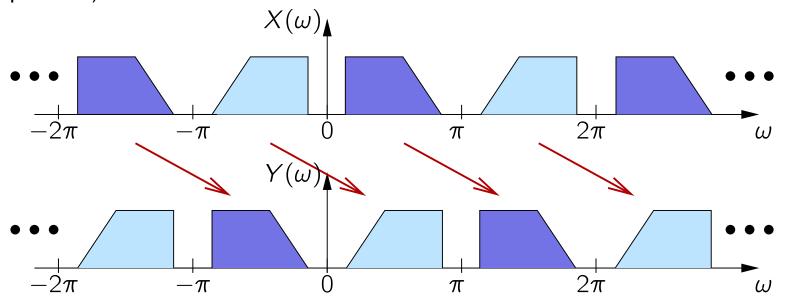
A simple scrambler swaps high and low frequencies in a signal



How can we carry out this operation? And reverse it?

### **Oplossing:**

First make a drawing of a larger part of the frequency domain (the spectrum is periodic)



It is seen that the scrambler shifts the signal over  $\pi$  rad:  $Y(\omega) = X(\omega - \pi)$ .

In time domain, a frequency shift corresponds to modulation:

$$Y(\omega) = X(\omega - \omega_0) \qquad \Leftrightarrow \qquad y[n] = x[n] \cdot e^{j\omega_0 n}$$

At  $\omega_0 = \pi$  we find  $y[n] = x[n] \cdot (-1)^n$ . Inverse operation: repeat the modulation.

### **Z-transform**

Determine the *z*-transform and the ROC of

a) 
$$x[n] = [\cdots, 0, 1, 2, 1, 0, \cdots]$$
  
b)  $x[n] = \operatorname{sgn}[n] = \begin{cases} 1, & n \ge 0, \\ -1, & n < 0 \end{cases}$ 

c)  $x[n] = -a^n u[-n]$ . For which values of *a* does the DTFT  $X(e^{j\omega})$  exist?

#### **Answer:**

a) 
$$X(z) = z + 2 + z^{-1}$$
, ROC:  $\mathbb{C} \setminus \{0, \infty\}$ .

b) For the causal part ( $n \ge 0$ ) we find

$$X_c(z) = \frac{1}{1 - z^{-1}}$$
 ROC:  $|z| > 1$ 

For the anti-causal part (assume n < 0):

$$X_{ac}(z) = -\frac{z}{1-z}$$
 ROC:  $|z| < 1$ 

The *z*-transform of the sum is (take the intersection of the ROC's)

$$X(z) = \frac{1}{1-z^{-1}} - \frac{z}{1-z}$$
 ROC: empty!

Hence the *z*-transform does not exist.

c) The signal exists for  $n \leq 0$ .

$$X(z) = -1 - \frac{z}{a} - \frac{z^2}{a^2} - \dots = -\frac{1}{1 - z/a} = \frac{az^{-1}}{1 - az^{-1}}$$
 ROC:  $|z| < a$ 

The DTFT exists if the ROC contains the unit circle: for |a| > 1.

### **Z-transform**

Determine the *z*-transform and the ROC of  $x_1[n] = u[n]$  and of  $x_2[n] = -u[-n-1]$ .

#### **Answer:**

$$X_1(z) = 1 + z^{-1} + z^{-2} + \dots = \frac{1}{1 - z^{-1}} \quad \text{ROC: } |z| > 1$$
$$X_2(z) = -z - z^2 - \dots = -\frac{z}{1 - z} = \frac{1}{1 - z^{-1}} \quad \text{ROC: } |z| < 1$$

The same *z*-transform but a different ROC.

Note the relation between the ROC (inside or outside of a circle) and (anti-)causality of x[n]!

#### **Inverse Z-transform**

Given  $X(z) = \frac{1 - z^{-10}}{1 - z^{-1}}$ , determine x[n]. (Assume that the ROC is |z| > 1.)

#### **Answer:**

For this ROC, x[n] is causal. Split X(z):

$$X(z) = \frac{1}{1 - z^{-1}} - \frac{z^{-10}}{1 - z^{-1}}$$

Using the *z*-transform of u[n] and a delay of 10 samples, we find

$$x[n] = u[n] - u[n - 10]$$

### **Difference equations and the** *z***-transform**

Given the 2nd order LTI system with difference equation

y[n] = 0.25y[n-2] + x[n]

Here, x[n] is the input and y[n] is the output.

- a) If  $y[n] = 0.5^n u[n]$ , then compute x[n].
- b) What is the system impulse response.

#### Answer

$$Y(z) = \frac{X(z)}{1 - 0.25z^{-2}}$$
 and  $Y(z) = \frac{1}{1 - 0.5z^{-1}}$ 

Hence

 $X(z) = 1 + 0.5z^{-1}$  so that  $x[n] = \delta[n] + 0.5\delta[n-1]$ 

The transfer function is (take X(z) = 1 so that Y(z) = H(z))

$$H(z) = \frac{1}{1 - 0.25z^{-2}} = 1 + 0.25z^{-2} + (0.25)^2 z^{-4} + \cdots$$

The impulse response is

$$h[n] = [1, 0, 0.25, 0, (0.25)^2, 0 \cdots]$$

This is also found if we carry out the recursion step-by-step taking as input an impulse  $x[n] = \delta[n]$ .

The impulse response is obtained in closed form by computing the poles of H(z), and splitting them (partial fraction expansion):

$$H(z) = \frac{1}{(1 - 0.5z^{-1})(1 + 0.5z^{-1})} = \frac{0.5}{1 - 0.5z^{-1}} + \frac{0.5}{1 + 0.5z^{-1}}$$

$$h[n] = 0.5 \{ (0.5)^n u[n] + (-0.5)^n u[n] \}$$

Verify that this gives the same result!

Determine the inverse *z*-transform of

$$X(z) = \frac{8 - 4z^{-1}}{z^{-2} + 6z^{-1} + 8}$$

Assume that x[n] is causal.

#### Answer

First write this as a polynomial in  $z^{-1}$  and make proper - already the case here.

Determine the poles and split into elementary terms (partial fraction expansion):

$$X(z) = \frac{2 - z^{-1}}{2(1 + 0.25z^{-1})(1 + 0.5z^{-1})} = \frac{A}{1 + 0.25z^{-1}} + \frac{B}{1 + 0.5z^{-1}}$$

where we find A = -3 en B = 4.

The inverse is found as

$$x[n] = -3(-0.25)^n u[n] + 4(-0.5)^n u[n]$$