

## EE2S11 SIGNALS AND SYSTEMS

Resit exam, 10 July 2023, 13:30–16:30

Closed book; one A4 (two sides) of handwritten notes permitted. No other tools except a basic pocket calculator permitted.

This exam consists of seven questions (31 points). Answer in Dutch or English. Make clear in your answer how you reach the final result; the road to the answer is very important. Write your name and student number on each answer sheet.

### Question 1 (4 points)

Determine the one-sided Laplace transform of the following signals and specify their ROC as well. *Note:*  $u(t)$  denotes the Heaviside unit step function.

(a)  $f(t) = \delta(2t)$

(b)  $g(t) = t u(t)$

(c)  $m(t) = (t - 1)u(t - 1)$

(d)  $n(t) = (t - 1)u(t)$

### Solution

1p (a)  $F(s) = 1/2$ , ROC =  $\mathbb{C}$

0.5p (b)  $G(s) = 1/s^2$ , ROC =  $\{s \in \mathbb{C}; \operatorname{Re}(s) > 0\}$

1p (c)  $M(s) = e^{-s}/s^2$ , ROC =  $\{s \in \mathbb{C}; \operatorname{Re}(s) > 0\}$

1.5p (d)  $N(s) = 1/s^2 - 1/s$ , ROC =  $\{s \in \mathbb{C}; \operatorname{Re}(s) > 0\}$

### Question 2 (4 points)

Find the inverse Laplace transform of the following signals:

(a)  $F(s) = \frac{4s + 20}{s^2 + 4s + 13}$ ,  $\operatorname{Re}(s) > -2$ .

(b)  $G(s) = \frac{1}{(s^2 + 4)(s^2 + 9)}$ ,  $\operatorname{Re}(s) > 0$ .

### Solution

2p (a)  $F(s) = \frac{4(s + 2) + 12}{(s + 2)^2 + 3^2} \longleftrightarrow f(t) = 4e^{-2t}(\cos 3t + \sin 3t)u(t)$

2p (b)  $G(s) = \frac{1}{5} \frac{1}{s^2 + 4} - \frac{1}{5} \frac{1}{s^2 + 9} \longleftrightarrow g(t) = \left( \frac{1}{10} \sin 2t - \frac{1}{15} \sin 3t \right) u(t)$

**Question 3 (6 points)**

Let  $x(t)$  be a periodic signal with period  $T_0 = 1$ . A single period of  $x(t)$  is denoted by  $x_1(t)$  and is given by

$$x_1(t) = \begin{cases} \sin(2\pi t) & 0 \leq t \leq 0.5 \\ 0 & 0.5 < t \leq 1. \end{cases}$$

- (a) Determine the one-sided Laplace transform of  $x_1(t)$  and its ROC.
- (b) Determine the DC component of the periodic signal  $x(t)$ .
- (c) Determine the Fourier coefficient  $X_1$  of the periodic signal  $x(t)$ .
- (d) Determine the Fourier coefficients  $X_k$  for  $k$  even and  $k \geq 2$ .
- (e) Determine the Fourier coefficients  $X_k$  for  $k$  odd and  $k \geq 2$ .

**Solution**

2p (a)  $X_1(s) = \frac{2\pi}{s^2 + 4\pi^2}(1 + e^{-0.5s}), \quad \text{ROC} = \mathbb{C}$

1p (b)  $X_0 = \frac{1}{\pi}$

1p (c)  $X_1 = \frac{1}{4j}$

1p (d)  $X_k = \frac{1}{\pi - k^2}, k \geq 2$  and  $k$  even

1p (e)  $X_k = 0, k \geq 2$  and  $k$  odd

**Question 4 (6 points)**

- (a) Given the signal  $x[n] = [\dots, 0, \boxed{3}, 2, 1, 0, 0, \dots]$ , where the ‘box’ denotes the value for  $n = 0$ . Determine  $r[n] = x[n] * x[-n]$  using the convolution sum.

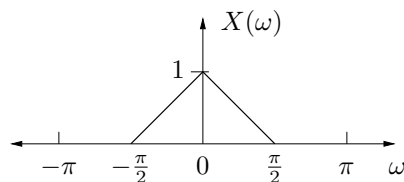
Determine the  $z$ -transform for the following discrete-time signal, and specify also the ROC:

(b)  $x[n] = (2)^n u[-n]$ ,

Determine the signal  $x[n]$  corresponding to the following  $z$ -transform:

(c)  $X(z) = \frac{2z^2}{(z-1)(z-2)}, \quad \text{ROC} = \{1 < |z| < 2\}$ .

The signal  $x[n]$  is specified by its DTFT (here we assume that  $X(\omega)$  is real-valued):



- (d) Determine and draw the DTFT of  $x_1[n] = x[n] \cos(\pi n/4)$ .

*Note:* specify the DTFT in terms of  $X(\omega)$ .

**Solution**

2p (a) Let  $y[n] = x[-n] = [\dots, 0, 1, 2, \boxed{3}, 0, 0, \dots]$ , then  $r[n] = \sum_{k=1}^2 x[k]y[n-k]$ ,

$$\begin{array}{r} k=0: \quad 3y[n]: \quad 3 \quad 6 \quad \boxed{9} \quad 0 \quad 0 \quad 0 \dots \\ k=1: \quad 2y[n-1]: \quad 0 \quad 2 \quad \boxed{4} \quad 6 \quad 0 \quad 0 \dots \\ k=2: \quad 1y[n-2]: \quad 0 \quad 0 \quad \boxed{1} \quad 2 \quad 3 \quad 0 \dots \\ \hline r[n]: \quad 3 \quad 8 \quad \boxed{14} \quad 8 \quad 3 \quad 0 \dots \end{array}$$

1p (b)  $X(z) = 1 + \frac{1}{2}z + \frac{1}{4}z^2 + \dots = \frac{1}{1 - \frac{1}{2}z}$ ; ROC =  $\{|z| < 2\}$ .

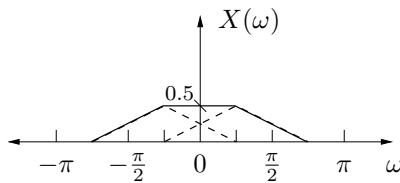
2p (c)

$$X(z) = \frac{-2z}{z-1} + \frac{4z}{z-2} = \underbrace{\frac{2}{1-z^{-1}}}_{\text{ROC: } |z| > 1} - \underbrace{\frac{2z}{1-\frac{1}{2}z}}_{\text{ROC: } |z| < 2}$$

The first term corresponds to a causal signal and the second to an anticausal signal. Hence

$$x[n] = 2u[n] - 2^{n+2}u[-n-1]$$

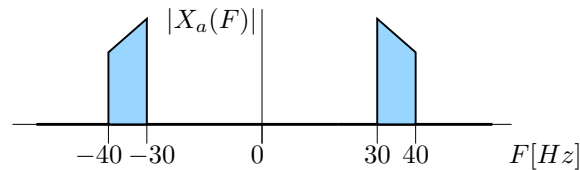
1p (d)  $X_1(\omega) = \frac{1}{2}X(\omega - \frac{\pi}{4}) + \frac{1}{2}X(\omega + \frac{\pi}{4})$ .



**Question 5 (3 points)**

A continuous-time signal  $x_a(t)$  has frequencies in the range 30 until 40 Hz. The signal is sampled with period  $T$  so that we obtain a time series  $x[n] = x_a(nT)$ .

The amplitude spectrum of  $x_a(t)$  appears as follows:



- (a) What is the Nyquist frequency at which we would have to sample to avoid any aliasing?
- (b) We sample the signal at 40 Hz. Make a drawing of the amplitude-spectrum  $|X(\omega)|$  resulting from this sample frequency. Also mark the frequencies.

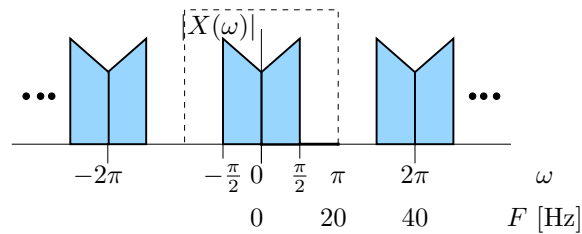
**Solution**

1p (a) Twice the highest frequency, 80 Hz.

2p (b) The spectrum is periodic with period  $F_s = 40$  Hz, take all shifts at multiples of 40 Hz and add these.

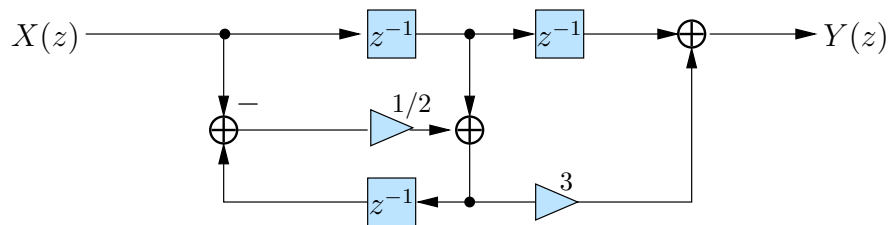
The component between 30 and 40 Hz also appears between  $-10$  and  $0$  Hz.

The component between  $-30$  and  $-40$  Hz also appears between  $0$  and  $10$  Hz. Further, the sample frequency  $F = 40$  Hz corresponds to  $\omega = 2\pi$ .



### Question 6 (3 points)

Given the realization



- Determine the transfer function  $H(z)$  corresponding to this realization.
- Is this a stable realization? (Why?)
- Is this a minimal realization? (Why?)

### Solution

2p (a) Introduce an extra parameter  $P(z)$  to the input of the multiplier “3”. We obtain

$$\begin{cases} P(z) = z^{-1}X(z) - \frac{1}{2}X(z) + \frac{1}{2}z^{-1}P(z) \\ Y(z) = z^{-2}X(z) + 3P(z) \end{cases}$$

An expression for  $P(z)$  is

$$P(z)\left(1 - \frac{1}{2}z^{-1}\right) = X(z)\left(z^{-1} - \frac{1}{2}\right)$$

$$P(z) = X(z)\frac{z^{-1} - \frac{1}{2}}{1 - \frac{1}{2}z^{-1}}$$

Eliminate  $P(z)$  in the expression for  $Y(z)$ , this results

$$\begin{aligned} Y(z) &= X(z)\left(z^{-2} + \frac{3(z^{-1} - \frac{1}{2})}{1 - \frac{1}{2}z^{-1}}\right) \\ &= X(z)\frac{-\frac{1}{2}z^{-3} + z^{-2} + 3z^{-1} - \frac{3}{2}}{1 - \frac{1}{2}z^{-1}} \end{aligned}$$

Hence

$$H(z) = \frac{-\frac{1}{2}z^{-3} + z^{-2} + 3z^{-1} - \frac{3}{2}}{1 - \frac{1}{2}z^{-1}}$$

0.5p (b) The pole is located at  $z = \frac{1}{2}$ , within the unit circle, hence stable.

0.5p (c) The highest degree (filter order) is 3, the number of delays is 3, hence minimal.

### Question 7 (5 points)

Design an analog 3rd order low-pass Butterworth filter  $H(\Omega)$  with a passband frequency of 2 rad/s, a stopband frequency of 4 rad/s and a maximal damping in the passband of 1 dB.

- (a) What is the general expression for the frequency response (squared-amplitude) of a prototype Butterworth low-pass filter.
- (b) Determine the unknown filter parameters such that the specifications are met.
- (c) For this filter, what is the minimal damping in the stopband ?
- (d) We wish to transform this filter to a *high*-pass filter  $G(\Omega)$  with a passband frequency  $\Omega'_p = 2$  rad/s. What transformation do we use, what is the resulting frequency response  $|G(\Omega)|^2$ , and what is the resulting stopband frequency?

### Solution

1p (a)  $|H(\Omega)|^2 = \frac{1}{1 + \epsilon^2(\Omega/\Omega_p)^{2N}}$ .

1p (b) At  $\Omega_p = 2$  we find  $|H(\Omega_p)|^2 = 10^{-1/10}$ .

$$\frac{1}{1 + \epsilon^2} = 10^{-1/10} \Rightarrow \epsilon = \sqrt{10^{1/10} - 1} = 0.5088$$

We already know that  $N = 3$ .

1p (c) At  $\Omega_s = 4$  we find

$$|H(\Omega_s)|^2 = \frac{1}{1 + \epsilon^2(4/2)^6} = 0.0569$$

This is  $-12.5$  dB.

2p (d)

$$\begin{aligned} \Omega &\rightarrow \frac{\Omega_p \Omega'_p}{\Omega} = \frac{4}{\Omega} \\ |G(\Omega)|^2 &= \frac{1}{1 + \epsilon^2(2/\Omega)^6} \\ \Omega'_s &= \frac{\Omega_p \Omega'_p}{\Omega_s} = \frac{4}{4} = 1 \end{aligned}$$